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Decentralised LQI-Type Load-frequency control with controlling delay of one sampling time for interconnected Power systems

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Abstract

This paper presents a new method of designing decentralised discrete-type load-frequency regulator with controlling delay of one sampling time for interconnected power systems. In this method, the interconnected multi-area electric energy system is decomposed into several subsystems, each of which is controlled separately by using a local feedback only. An especially attractive feature of the proposed control scheme is that it considers the time delay due to the computation time of the control law and the transmission time of the system data over the telemeter links to the controlling plant. An additional feature is that the construction of the regulator is based on the conventional tie-line bias control. The proposed control scheme is illustrated by digital simulation of a 2-area system provided with reheat steam turbines. The results show that this discrete-type regulator can act satisfactorily for improving dynamic responses of the load-frequency control.

1. Introduction

The load-frequency control (LFC) problem has been one of the major subjects concerning power-system engineers, and the objective of LFC is to minimize the transient errors in the frequency and the scheduled tie-line power and ensure zero steady-state errors of these two quantities¹⁻³. For many years, a considerable research effort has been devoted to the development of control strategies for the LFC problem using continuous-time optimization techniques³⁻⁶.

However, further consideration may be required for practical implementation of control strategies designed by using continuous-time optimization techniques, because, in practical power systems, the system data (frequency, tie-line power etc.) are available in the discrete form, i.e. the system data are first sampled and then transferred over telemeter links. Also the use of digital computers has become practically indispensable to electric energy systems, because, with the use of digital computers, it is now possible, through suitable software, to realise a wide range of control strategies. Accordingly, it may be necessary to construct discrete-type regulators for such practical situations. However, until now, few works⁷⁻⁹ have been done concerning discrete-time load-frequency control, and in most of the past works, they have ignored the time delay due to the computational time of the control law and the transmission time of the system data over the telemeter links to the controlling plant.

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Furthermore, it is recognized that the implementation of a centralised load-frequency control possesses certain difficulties, when the size and complexity of the interconnected power systems increases. Specifically, these difficulties can be traced to the need for elaborate instrumentation and telemetry of the required data to the central controller. Also the computational requirements grow very fast with the number of interconnected areas. In recent years, significant efforts have been made to establish suitable decentralised regulator for large interconnected power systems¹⁰⁻¹¹.

This paper presents a new method of designing decentralised discrete-type load-frequency regulator with controlling delay of one sampling time for interconnected power systems. In this method, the interconnected multi-area electric energy system is decomposed into several subsystems, each of which is controlled separately by using a local feedback only. An especially attractive feature of the proposed control scheme is that it considers the time delay due to the computation time of the control law and the transmission time of the system data over the telemeter links to the controlling plant. An additional feature is that the construction of the regulator is based on the conventional tie-line bias control, which is used by most utilities of the present day.

In this paper, we apply the newly designed regulator to 2-area power system provided with reheat steam turbines. The results show that the proposed discrete-type regulator can act satisfactorily for improving dynamic responses of the load-frequency control.

2. Notation

- f^* = nominal system frequency
- i = subscript referring to area i
- Δf_i = incremental frequency deviation
- ΔP_{ui} = incremental generation change
- ΔP_{ri} = incremental generation change during steam reheat
- ΔP_{gi} = incremental change in governor valve position
- ΔP_{tiei} = incremental change in tie-line power
- ΔP_{ci} = incremental change in speed changer position
- ΔP_{di} = incremental load demand change
- H_i = inertia constant
- D_i = load frequency constant
- k_{hi} = high pressure turbine power fraction
- T_{ri} = reheat time constant
- T_{ij} = synchronising coefficient
- T_{ui} = steam chest time constant
- T_{gi} = speed governor time constant
- R_i = self regulation parameter for the governor
- β_i = frequency bias parameter
- T_s = sampling period
- Δ = small deviation of state variable
- s = Laplace operator

(Other symbols are defined in the text.)

3. Problem formulation

A typical 2-area power system with reheat steam turbines is shown in Figure 1 for LFC. In state variable form, the continuous-time dynamics of the i-th controlling plant in an n-interconnected system is described by a set of linear differential equations with input delays:

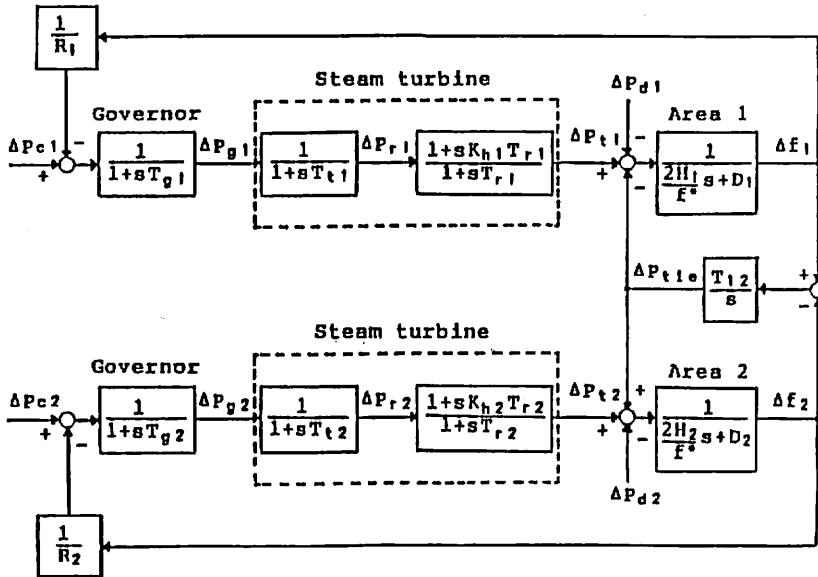


Fig. 1 Block diagram of 2-area reheat thermal system

$$\dot{x}_i(t) = A_i x_i(t) + \sum_j A_{ij} x_j(t) + B_i u_i(t - T_s) + \Gamma_i v_i(t) \quad x_i(0) = 0 \quad (1)$$

where

$$x_i = [\Delta f_i \ \Delta P_{ti} \ \Delta P_{ri} \ \Delta P_{gi} \ \Delta P_{tiei}]^T \quad u_i = [\Delta P_{ci}] \quad v_i = [\Delta P_{di}]$$

and T_s is the sampling period. In this case, $u_i(t - T_s)$ with controlling delay of one sampling time is sufficient to consider the effect of the time delay due to the computation time of the control law and the transmission time of the system data over the telemeter links to the controlling plant for practical sampling periods ($1.5 \leq T_s \leq 2.5$) used for LFC. The matrices in equation (1) are given by:

$$\begin{aligned}
 A_i = & \begin{bmatrix} -\frac{f^* D_i}{2H_i} & \frac{f^*}{2H_i} & 0 & 0 & -\frac{f^*}{2H_i} \\ 0 & -\frac{1}{T_{r1}} & \frac{1}{T_{r1}} - \frac{k_{hl}}{T_{ti}} & \frac{k_{hl}}{T_{ti}} & 0 \\ 0 & 0 & -\frac{1}{T_{ti}} & \frac{1}{T_{ti}} & 0 \\ -\frac{1}{T_{g1} R_i} & 0 & 0 & -\frac{1}{T_{g1}} & 0 \\ \sum_j T_{ij} & 0 & 0 & 0 & 0 \end{bmatrix} & B_i = \begin{bmatrix} 0 \\ 0 \\ 0 \\ \frac{1}{T_{g1}} \\ 0 \end{bmatrix} \\
 A_{ij} = & \begin{bmatrix} 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ -T_{ij} & 0 & 0 & 0 & 0 \end{bmatrix} & \Gamma_i = \begin{bmatrix} -\frac{f^*}{2H_i} \\ 0 \\ 0 \\ 0 \\ 0 \end{bmatrix}
 \end{aligned} \tag{2}$$

However, if it is assumed that $v_i(t)$ represents known disturbances, equation (1) can be rewritten in the standard state form as

$$\dot{x}_i(t) = A_i x_i(t) + B_i u_i(t - T_s) \quad x_i(0) = -x_{is} \tag{3}$$

where x_i and u_i are the new state and control vectors which equal the old vectors minus these steady state values x_{is} and u_{is} . In equation (3), the coupling terms between the areas are set equal to zero to simplify the problem formulation (non-zero coupling is considered in the example).

Since the control signals $u_i(t)$, ($i=1, 2, \dots, n$) are now outputs of sample-and-hold devices, they are constant during the sampling period T_s and described by

$$u_i(\overline{(k-1)T_s}) \quad \text{for } kT_s \leq t < \overline{(k+1)T_s} \tag{4}$$

Thus the solution of equation (3) is found to be

$$\begin{aligned}
 x_i(t) &= e^{A_i(t-kT_s)} x_i(kT_s) + \int_{kT_s}^t e^{A_i(t-\tau)} B_i d\tau u_i(\overline{(k-1)T_s}) \\
 &= e^{A_i(t-kT_s)} x_i(kT_s) + [e^{A_i(t-kT_s)} - I] A_i^{-1} B_i u_i(\overline{(k-1)T_s})
 \end{aligned} \tag{5}$$

The equation can be further modified to describe the transition of the states of the digital system at the sampling instants only. By letting $t = \overline{(k+1)T_s}$, equation (5) becomes

$$x_i(\overline{(k+1)T_s}) = \Phi_i x_i(kT_s) + \Psi_i u_i(\overline{(k-1)T_s}) \tag{6}$$

where

$$\Phi_i = e^{A_i T_s} \quad \Psi_i = [e^{A_i T_s} - I] A_i^{-1} B_i$$

I = identity matrix

Since the steady-state errors of frequency and tie-line power deviations should be driven to zero by LFC, we suggest as feedback signals accumulative quantities of the area control error on the basis of linear quadratic integrating technique¹². They can be defined as follows:

$$\tilde{x}_i(\overline{k+1}T_s) = \sum_{m=0}^{\overline{k+1}T_s} \{ \beta_i \Delta f_i(m) + \Delta P_{tie}(m) \} \tag{7}$$

Using equation (6), equation (7) can be rewritten as

$$\begin{aligned} \tilde{x}_i(\overline{k+1}T_s) &= \tilde{x}_i(kT_s) + C_i x_i(\overline{k+1}T_s) \\ &= \tilde{x}_i(kT_s) + C_i \Phi_i x_i(kT_s) + C_i \Psi_i u_i(\overline{k-1}T_s) \end{aligned} \tag{8}$$

where

$$C_i = [\beta_i \quad 0 \quad 0 \quad 0 \quad 1]$$

Now, define the augmented state vector as

$$\hat{x}_i(k)^T = [x_i(k)^T \quad \tilde{x}_i(k) \quad u_i(k-1)] \tag{9}$$

where $x_i(k)$, $\tilde{x}_i(k)$ and $u_i(k-1)$ imply $x_i(kT_s)$, $\tilde{x}_i(kT_s)$ and $u_i(k-1T_s)$, respectively.

The augmented set of difference equations for LFC system are

$$\hat{x}_i(k+1) = \hat{\Phi}_i \hat{x}_i(k) + \hat{\Psi}_i u_i(k) \tag{10}$$

where

$$\hat{\Phi}_i = \begin{bmatrix} \Phi_i & 0 & \Psi_i \\ C_i \Phi_i & 1 & C_i \Psi_i \\ 0 & 0 & 0 \end{bmatrix} \quad \hat{\Psi}_i = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$

An optimal control can be achieved by minimizing a quadratic cost function of the form

$$J_i = E \left[\sum_{k=0}^{\infty} \{ \hat{x}_i(k)^T Q_i \hat{x}_i(k) + u_i(k)^T R_i u_i(k) \} \right] \tag{11}$$

subject to the system control constraint of equation (10), where E is the expectation operator over the random initial state, Q_i is 7×7 symmetric positive-semi-definite constant matrix and R_i is positive scalar value.

Now, define the output vector $y_i(k)$ as

$$y_i(k) = [\Delta f_i(k) \quad \Delta P_{tie}(k)]^T \tag{12}$$

These variables are easily measurable variables and used as the feedback signals of the conventional tie-line bias control. Additional, $\tilde{x}_i(k)$ constructed by the elements of equation (12) and $u_i(k-1)$ are measurable variables. Therefore, suppose that control

variable $u_i(k)$ is represented by a linear combination of measurable variables $y_i(k)$, $\tilde{x}_i(k)$ and $u_i(k-1)$ as follows:

$$u_i(k) = -F_1 \hat{D}_1 \hat{x}_i(k) \quad (13)$$

where

$$F_1^\top = \begin{bmatrix} F_{11}^\top \\ F_{12} \\ F_{13} \end{bmatrix} \quad D_1 = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 \end{bmatrix} \quad \hat{D}_1 = \begin{bmatrix} D_1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

and F_1 is undefined feedback gain vector.

Substituting for $u_i(k)$ from equation (13) in equation (10), the closed-loop system matrix becomes $\Gamma_1 = \hat{\Phi}_1 - \hat{\Psi}_1 F_1 \hat{D}_1$, equation (10) reduces to

$$\hat{x}_i(k+1) = \Gamma_1 \hat{x}_i(k) \quad (14)$$

and the performance index J_1 is also expressed as

$$J_1 = E [\hat{x}_i(0)^\top P_1 \hat{x}_i(0)] \quad (15)$$

where P_1 is the solution of the Lyapunov Matrix equation

$$P_1 - \Gamma_1^\top P_1 \Gamma_1 = Q_1 + \hat{D}_1^\top F_1^\top R_1 F_1 \hat{D}_1 \quad (16)$$

On assuming now that $\hat{x}_i(0)$ is uniformly distributed over the surface of a hypersphere¹³, the performance index becomes

$$J_1 = \text{tr} P_1 \quad (17)$$

where $\text{tr} P_1$ is the sum of the diagonal terms of P_1 .

Therefore F_1 for the optimal controller must be determined in such a manner that J_1 can be minimized subject to the constraint given by equation (16). For minimizing J_1 , the Hamiltonian

$$H_1 = \text{tr} P_1 + \text{tr} L_1 (Q_1 + \hat{D}_1^\top F_1^\top R_1 F_1 \hat{D}_1 + \Gamma_1^\top P_1 \Gamma_1 - P_1) \quad (18)$$

is chosen and the necessary conditions for minimization of H_1

$$\frac{\partial H_1}{\partial F_1} = 0, \quad \frac{\partial H_1}{\partial L_1} = 0, \quad \frac{\partial H_1}{\partial P_1} = 0, \quad (19)$$

yield the following solutions:

$$F_1 = (R_1 + \hat{\Psi}_1^\top P_1 \hat{\Psi}_1)^{-1} \hat{\Psi}_1^\top P_1 \hat{\Phi}_1 L_1 \hat{D}_1^\top (\hat{D}_1 L_1 \hat{D}_1^\top)^{-1} \quad (20)$$

$$(\hat{\Phi}_1 - \hat{\Psi}_1 F_1 \hat{D}_1)^\top P_1 (\hat{\Phi}_1 - \hat{\Psi}_1 F_1 \hat{D}_1) - P_1 + Q_1 + \hat{D}_1 F_1^\top R_1 F_1 \hat{D}_1 = 0 \quad (21)$$

$$(\hat{\Phi}_1 - \hat{\Psi}_1 F_1 \hat{D}_1) L_1 (\hat{\Phi}_1 - \hat{\Psi}_1 F_1 \hat{D}_1)^\top - L_1 + I = 0 \quad (22)$$

The equations (20),(21) and (22) are solved iteratively in the following steps:

Step 1: Assume an initial value of F_1, \hat{D}_1 , such that all eigenvalues of $(\hat{\Phi}_1 - \hat{\Psi}_1 F_1 \hat{D}_1)$ lie inside of the unite circle in the z-plane.

Step 2: Use this value of F_1, \hat{D}_1 to solve equations (21) and (22) for P_1 and L_1 .

Step 3: Find a new value of F_1 by substituting these values of P_1 and L_1 in equation (20).

Step 4: Obtain a new value of F_1, \hat{D}_1 by using the value of F_1 from Step 3, and then return to Step 2.

By executing the above steps, we can uniquely determine the optimal feedback gain vector F_1 .

4. Application

A 2-area power system with reheat steam turbines as shown in Figure 1 is used to evaluate the effectiveness of the proposed load-frequency controller. The values of the system parameters are given in Table 1. The weighting matrices in equation (11) are shown as $Q_i = I$ and $R_i = 1$, ($i=1, 2$), respectively. The optimal feedback gain vectors $F_1, (i=1, 2)$ are obtained as the solution of equations (20),(21) and (22). The results are indicated in Table 2 ($\Delta ACE_1(m)$ implies $\beta_1 \Delta f_1(m) + \Delta P_{tie1}(m)$). Figures 2-5 indicate the control effects achived by the proposed load-frequency regulators under a step-load change $\Delta P_{d1} = 0.01$ pu MW in area 1. Curves in Figures 2-5 show the performance of $\Delta f_1, \Delta f_2, \Delta P_{tie}$ and ΔP_c for the optimal system and the uncontrolled system (here set $\Delta P_{tie} = \Delta P_{tie1} = -\Delta P_{tie2}$). The solid curves show the performance of the optimal system, whereas the dotted curves show the performance of the uncontrolled system. It can be seen from these figures that the transient errors in the frequency and the scheduled tie-line power are much reduced and zero steady-state errors of these quantities are ensured, by the regulator with the optimal

Table 1 System Parameters.

$H_1 = 5s$	$D_1 = 8.33 \times 10^{-3} \text{puMW/Hz}$
$R_1 = 2.4 \text{Hz/puMW}$	$T_{t1} = 0.3s$
$T_{g1} = 0.08s$	$T_{r1} = 10.0s$
$K_{h1} = 0.5$	$\beta_1 = D_1 + 1/R_1 = 0.425 \text{puMW/Hz}$
	($i=1,2$)
$f^* = 60 \text{Hz}$	$T_{12} = 0.545 \text{puMW/Hz} \cdot s$
$T_s = 2.0s$	

Table 2 Optimal feedback gain vector F_1

Ourput feedback	F_1	Output feedback	F_2
$\Delta f_1(k)$	-0.1776	$\Delta f_2(k)$	-0.1776
$\Delta P_{tie1}(k)$	0.3594	$\Delta P_{tie2}(k)$	0.3594
$\sum_{m=0}^k \Delta ACE_1(m)$	0.5320	$\sum_{m=0}^k \Delta ACE_2(m)$	0.5320
$\Delta P_{c1}(k-1)$	0.1052	$\Delta P_{c2}(k-1)$	0.1052

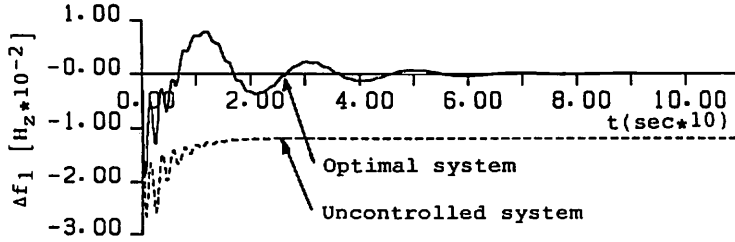


Fig. 2 Responses of Δf_1

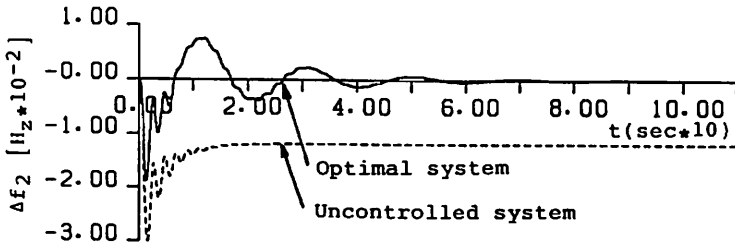


Fig. 3 Responses of Δf_2

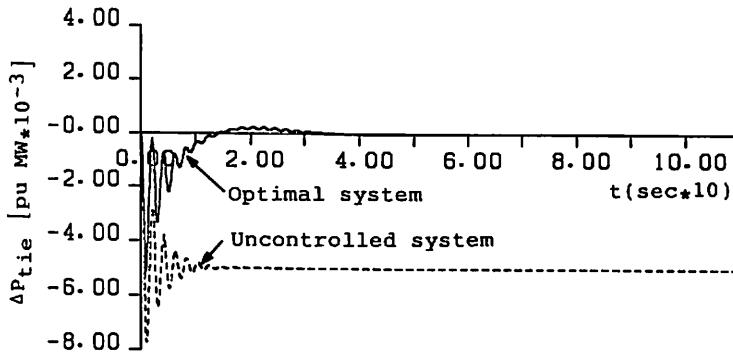


Fig. 4 Responses of ΔP_{tie}

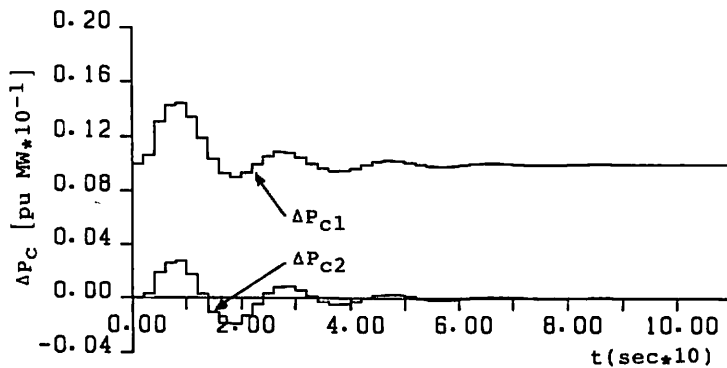


Fig. 5 Responses of ΔP_{c1} and ΔP_{c2}

gains. It may thus be inferred that the proposed control scheme can be acceptable for LFC of interconnected power systems.

5. Conclusions

In this paper, a new method of designing a decentralised discrete-type load-frequency regulator with controlling delay of one sampling time has been proposed, and the effectiveness of the proposed regulator has been shown in 2-area power system with reheat steam turbines. The major contributions of this paper are as follows:

(a) The proposed control scheme is very realistic, because the regulator is designed to consider the effect of the time delay due to the computation time of the control law and the transmission time of the system data over the telemeter links to the controlling plant.

(b) The realisation of this regulator is very easy, because the feedback data, required to form the control signals are only sampled frequency and tie-line power, which are used by most of the utilities of the present day.

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